

EEE443 Digital Signal Processing  
Implementation of Discrete-Time  
Systems

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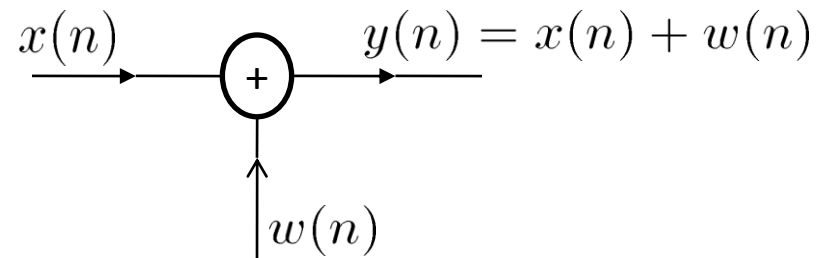
# Introduction

- A linear-time invariant system (LTI) is described by linear constant coefficient difference equation (LCCDE)
  - If we are given a rational transfer function having the input  $x(n)$  and output  $y(n)$ , the LTI is described as
$$y(n) = \sum_{k=0}^q b(k)x(n-k) - \sum_{k=1}^p a(k)y(n-k)$$
  - What we understand from this equation?
    - Recursive system – IIR (which have the terms of  $a(k)$ )
    - Non-recursive system – FIR (which its all  $a(k) = 0$ )

# Block Diagram for Illustration Purpose

- There are three basic computational elements:

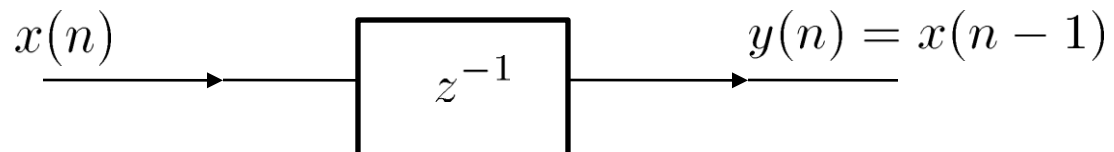
– Adders



– Multipliers



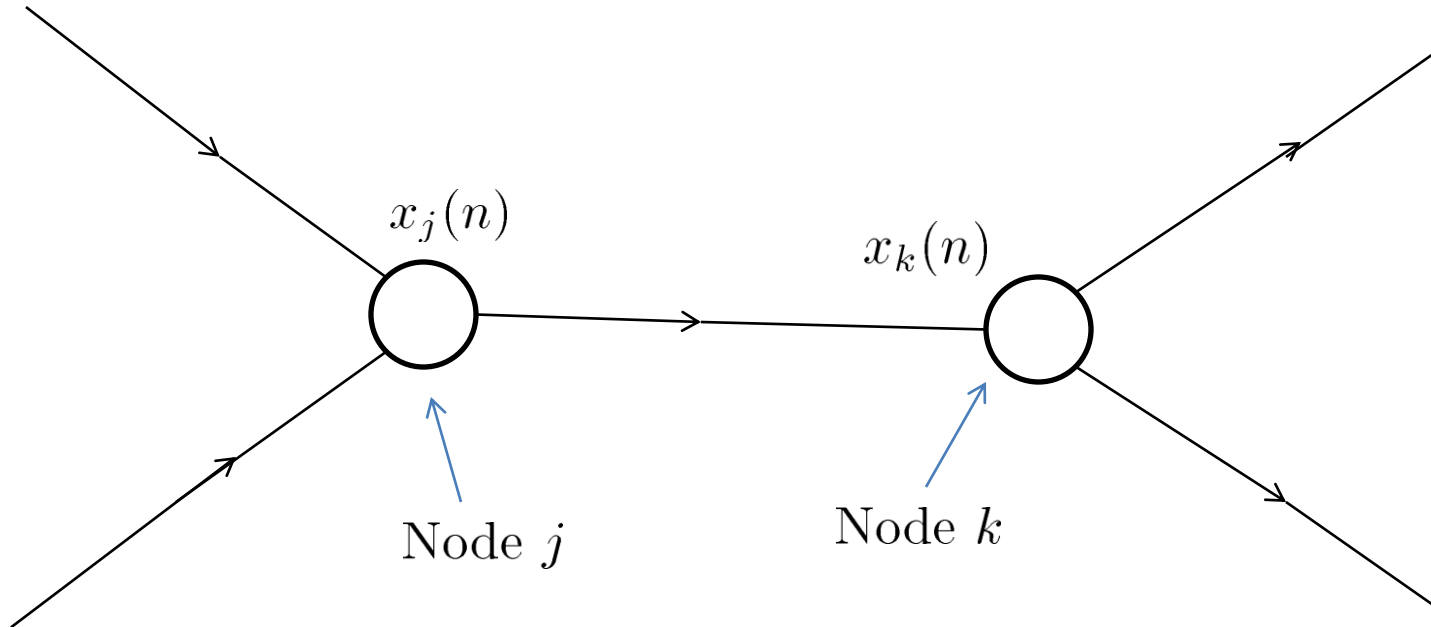
– Delays



# Signal Flow Graph

- Originated from the basic computational elements
- Signal flow graph is drawn using nodes to show the network of directed branches
- The nodes correspond to two type of points
  - Adders – more than one incoming branches
  - Branch points – more than one outcoming branches

# Signal Flow Graph Example



# Facts About Signal Flow Graph

- Output of each branch is a linear transformation of the branch input and the linear operator is indicated next to the arrow
  - Notice that for these linear operators, multipliers and delays will be included
- Two types of special nodes:
  - Source nodes – these are nodes that have no incoming branches and are used for sequences that input to the filter
  - Sink nodes – these are nodes that have only entering branches and are used to represent output sequences

# Structures of FIR Systems

- FIR – finite impulse response
- What we know about FIR?
  - All its  $a(k)$  terms will be zero
  - Non-recursive
- A causal FIR filter/system has a transfer function:

$$H(z) = \sum_{n=0}^{M-1} h(n)z^{-n}$$

- Notice that this transfer function has no poles (non-recursive)
- An FIR system is described by the difference equation

$$y(n) = \sum_{k=0}^{M-1} b_k x(n-k) \quad \xrightarrow{\text{System Function}} \quad H(z) = \sum_{k=0}^{M-1} b_k z^{-k}$$

- There are several different realization block diagrams for this system:
  - Direct form, cascade form, frequency-sampling, and lattice

# Structures of FIR Systems (cnt'd)

- Unit sample response of the FIR system is identical to the coefficient  $\{b_k\}$ , ie.

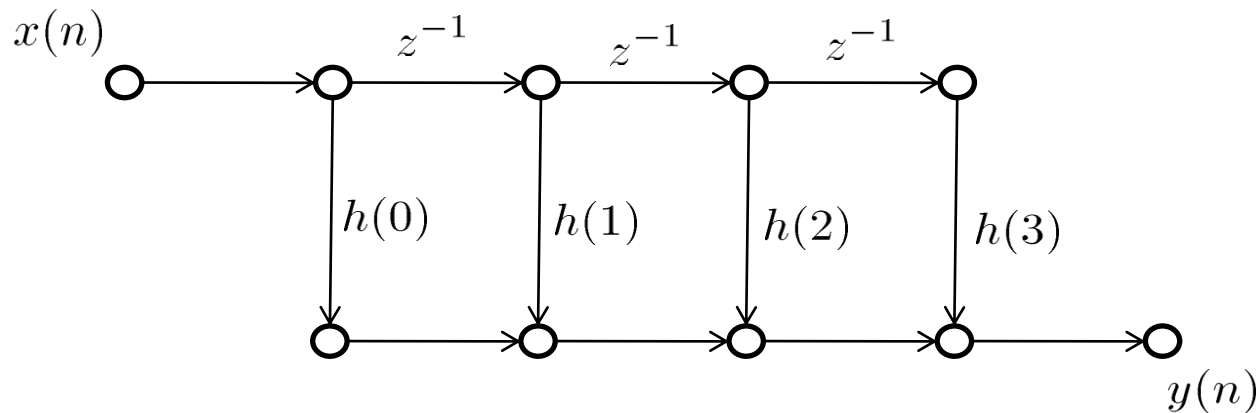
$$h(n) = \begin{cases} b_n, & 0 \leq n \leq M - 1 \\ 0, & \text{otherwise} \end{cases}$$

- $M$  is the length of the FIR filter



# Direct Form (1)

- The most common way to implement an FIR
- Called a tapped delay line or a transversal system (below is shown the signal flow graph)



# Direct Form (2)

$$y(n) = \sum_{k=0}^{M-1} h(k)x(n-k)$$

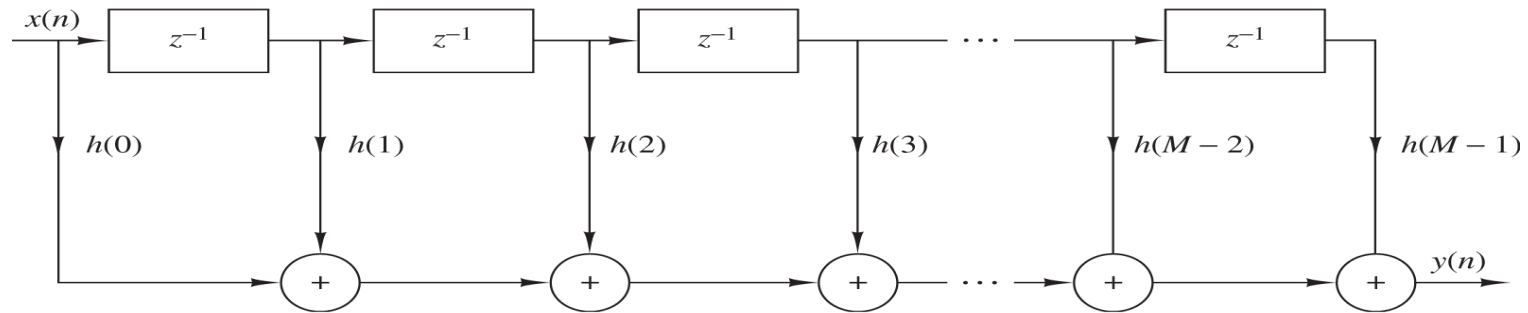


Figure 9.2.1 Direct-form realization of FIR system.

- This structure requires
  - M-1 memory locations for storing the M-1 previous inputs
  - M multiplications
  - M-1 additions per output point
- When the FIR system has linear phase, the unit sample response of the system satisfies either the symmetry or asymmetry condition

$$h(n) = \pm h(M-1-n)$$

- The advantage  $\rightarrow$  number of multiplications is reduced:
  - From M to M/2 for M even
  - From M to (M-1)/2 for M odd

# Cascade Form (1)

- The transfer function is factored into a product of first order factors

$$H(z) = \sum_{n=0}^N h(n)z^{-n} = A \prod_{k=1}^N (a - \alpha_k z^{-1}) \quad \text{where, for } \alpha_k, k = 1, 2, \dots$$

- If  $h(n)$  is real, the complex roots of  $H(z)$  occur in complex conjugate pairs, and therefore these conjugate pairs may be combined to form second order factors with real coefficients

$$H(z) = A \prod_{k=1}^N [1 + b_k(1)z^{-1} + b_k(2)z^{-2}]$$

# Cascade Form (2)

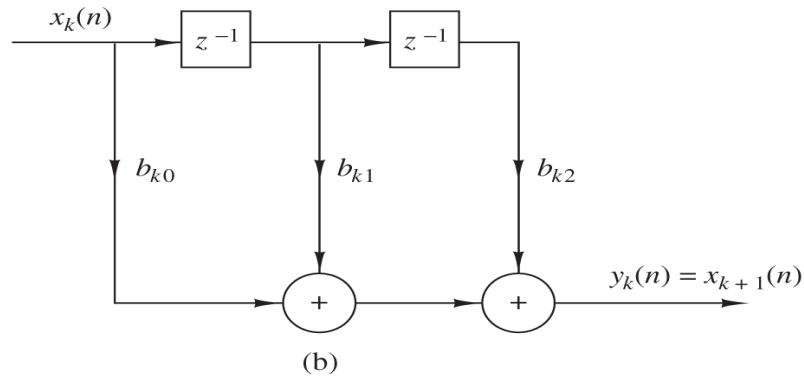
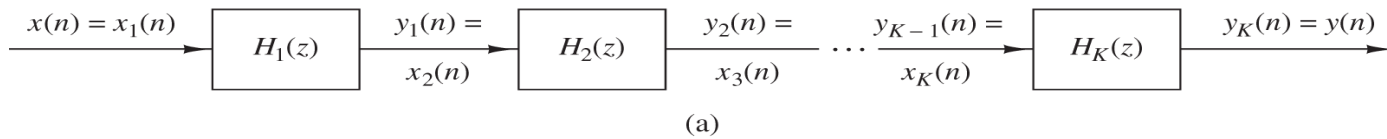
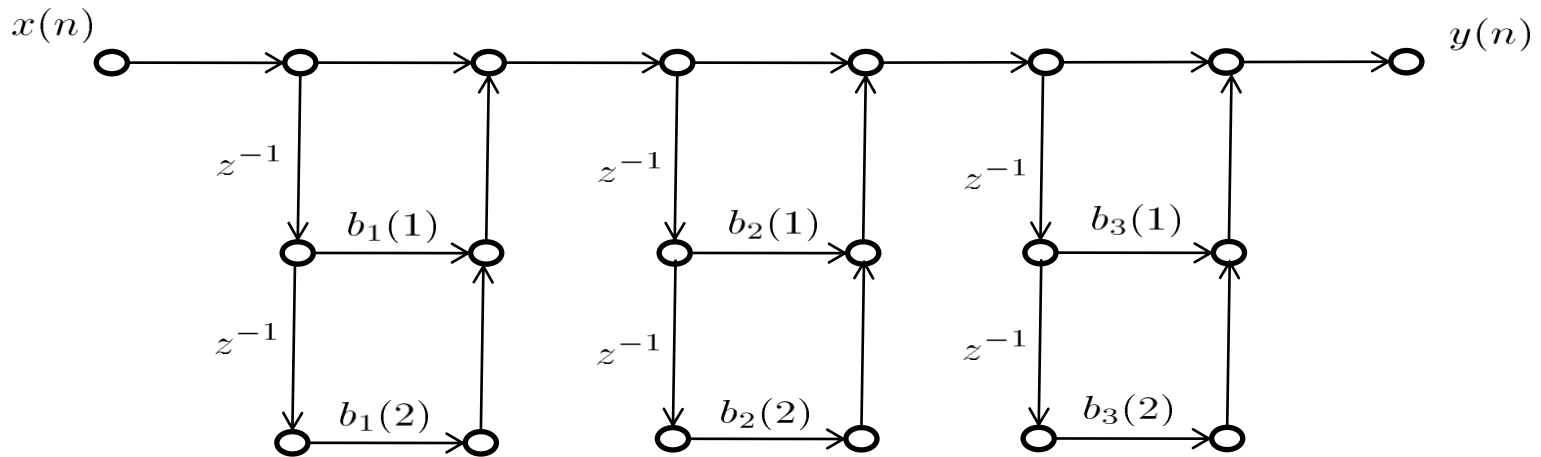


Figure 9.2.3 Cascade realization of an FIR system.

# Linear Phase Filter/Systems

- A linear phase filter/system is said to have linear phase if the frequency response has the form

$$H(e^{j\omega}) = |H(e^{j\omega})|e^{-j\alpha\omega}$$

- The constant group delay  $\tau_h(\omega) = \alpha$
- The unit impulse response of a linear phase system is either (taking into account  $N = M - 1$ )
  - Symmetric  $h(n) = h(N - n)$
  - Non-symmetric  $h(n) = -h(N - n)$

# Type of FIR Filter with Linear Phase

- The symmetry can be exploited to simplify the system/filter structure
- Therefore, there are four types of FIR filter
  - Type I :  $N$  is even and  $h(n)$  is symmetric
  - Type II :  $N$  is odd and  $h(n)$  is symmetric
  - Type III:  $N$  is even and  $h(n)$  is antisymmetric
  - Type IV:  $N$  is odd and  $h(n)$  is antisymmetric

# Structures of IIR Filters

- As being described before, a causal IIR filter with rational system or transfer function is expressed as

$$H(z) = \frac{B(z)}{A(z)} = \frac{\sum_{k=0}^q b(k)z^{-k}}{1 + \sum_{k=1}^p a(k)z^{-k}}$$

- Describing it in LCCDE

$$y(n) = \sum_{k=0}^q b(k)x(n-k) - \sum_{k=1}^p a(k)y(n-k)$$

- Similar to FIR, there are four types of IIR structures or realizations
  - Direct form, cascade form, lattice, and lattice-ladder

# Direct-Form (1)

- The rational system function that characterizes an IIR system can be viewed as two systems in cascade

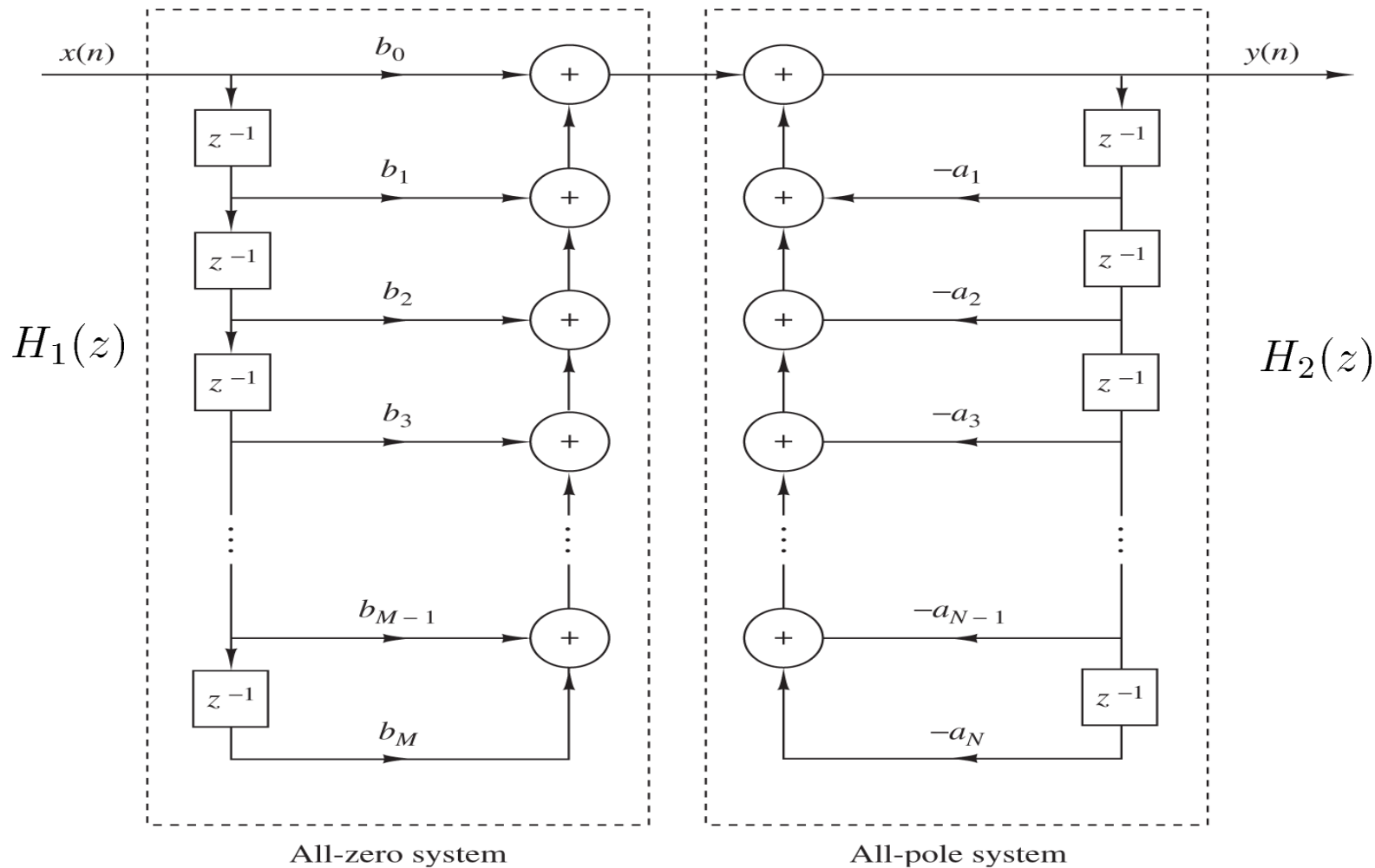
$$H(z) = H_1(z)H_2(z)$$

$$H_1(z) = \sum_{k=0}^M b_k z^{-k} \quad H_2(z) = \frac{1}{1 + \sum_{k=1}^N a_k z^{-k}}$$

- In Section 2.5, we have come across this where the cases were  $H_1(z)$  precedes  $H_2(z)$ , or vice versa
- Since the top equation is an FIR system, we know how write the diagram for this (Figure 9.2.1)
- To finalize the structure for an IIR, we attach the all-pole system in cascade with all-zero system (FIR system) as shown in Figure 9.3.1



# Direct-Form (2)



**Figure 9.3.1** Direct form I realization.

# Direct-Form (3)

- Figure 9.3.1 can be redrawn as shown in Figure 9.3.2 by considering linear equation for all-pole filter

$$w(n) = - \sum_{k=1}^N a_k w(n - k) + x(n)$$

- Since  $w(n)$  is the input to the all-zero system, its output is

$$y(n) = \sum_{k=0}^M b_k w(n - k)$$

- Looking at both equations above, we may note that both involve delayed versions of the sequence  $\{w(n)\}$ 
  - Resulting only a single delay line or single set of memory locations is required for storing the past values of  $\{w(n)\}$
  - This structure type is called a direct form II
  - Requires  $M+N+1$  multiplications,  $M+N$  additions and max of  $\{M,N\}$  memory locations
  - Is said to be *canonic*  $\rightarrow$  as it minimizes the number of memory locations
- “Direct-form” is given due to it can be obtained directly from the rational function or system function
  - BUT SENSITIVE TO PARAMETER QUANTIZATION  $\rightarrow$  NOT RECOMMENDED IN PRACTICAL APPLICATIONS

# Direct-Form (4)

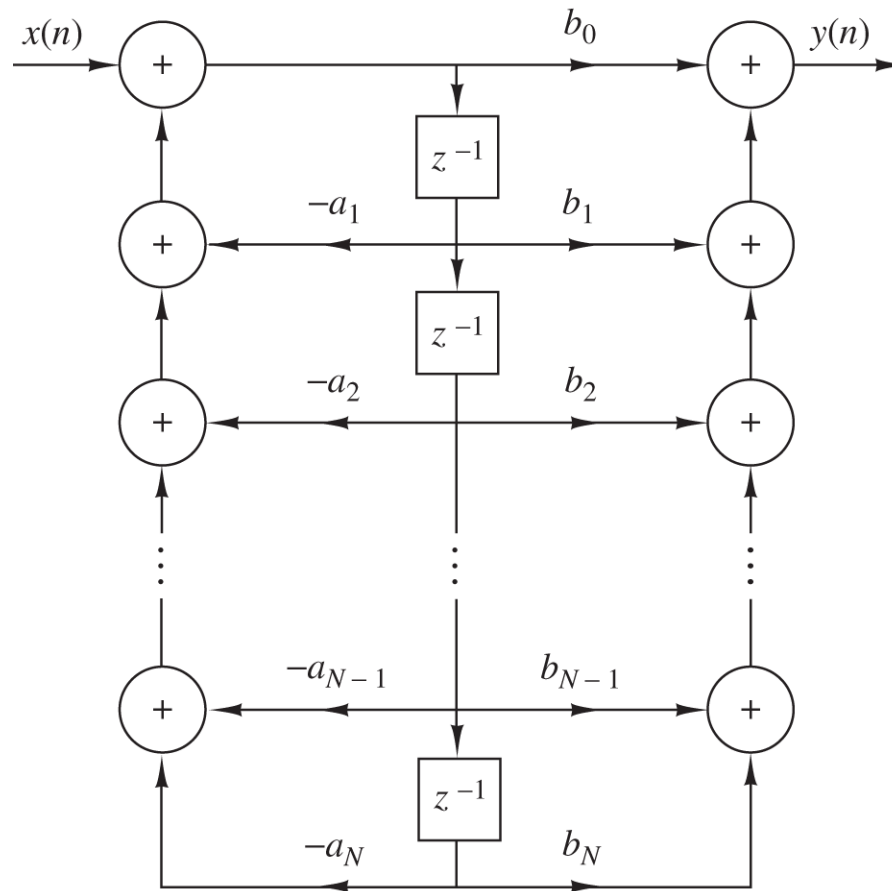
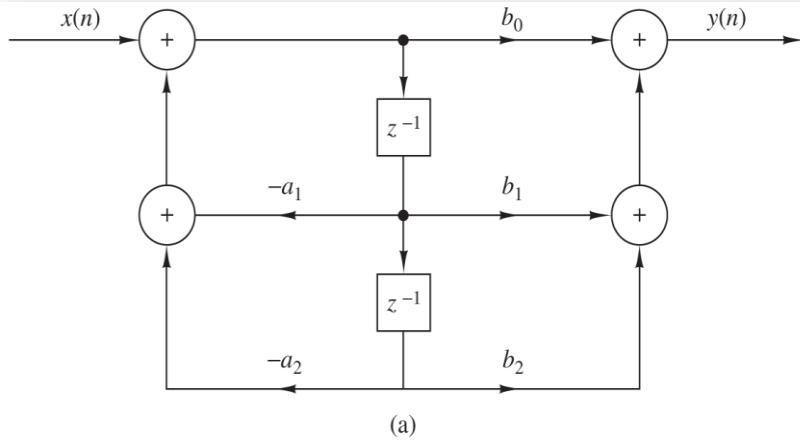


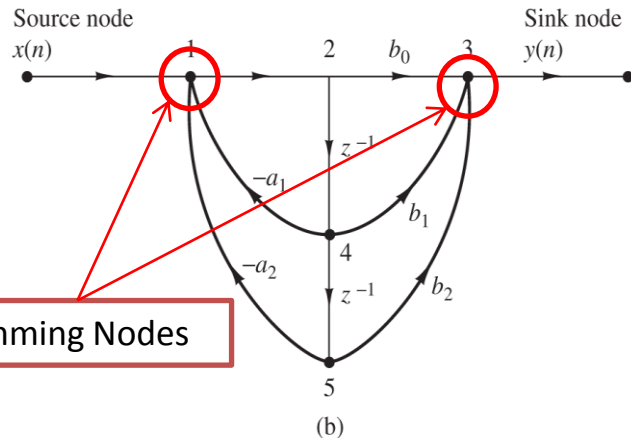
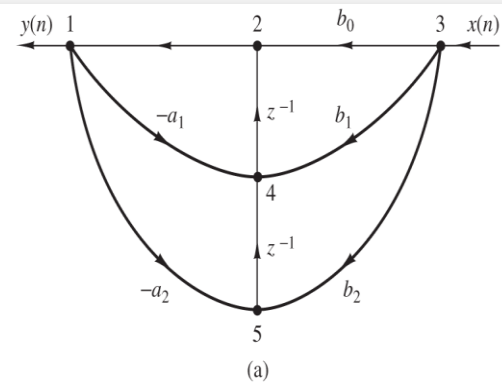
Figure 9.3.2 Direct form II realization ( $N = M$ ).

# About Signal Flow Graphs and Transposed Structures

Signal Flow Graph



Transposed Structures



Summing Nodes

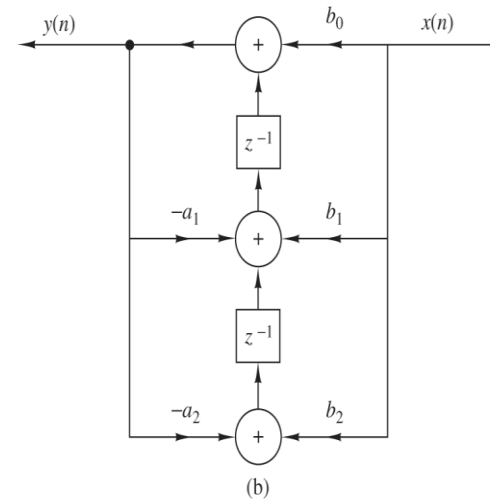


Figure 9.3.3 Second-order filter structure (a) and its signal flow graph (b).

Figure 9.3.4 Signal flow graph of transposed structure (a) and its realization (b).

# Let's try using the transposition theorem to the direct form II

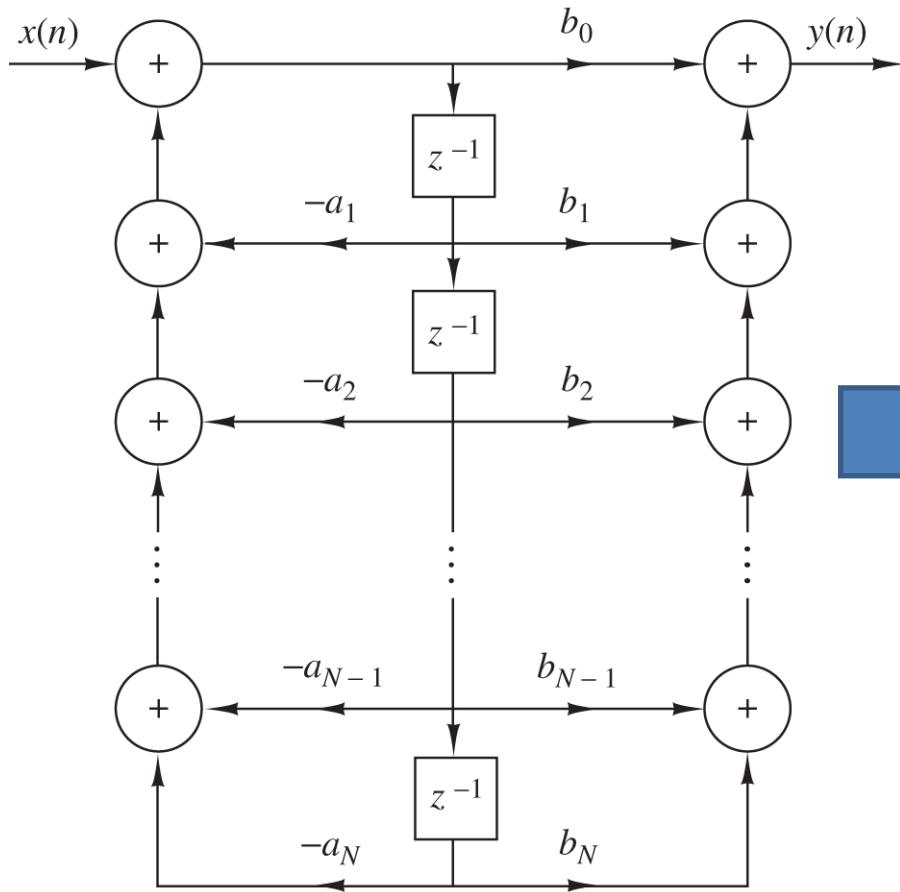


Figure 9.3.2 Direct form II realization ( $N = M$ ).

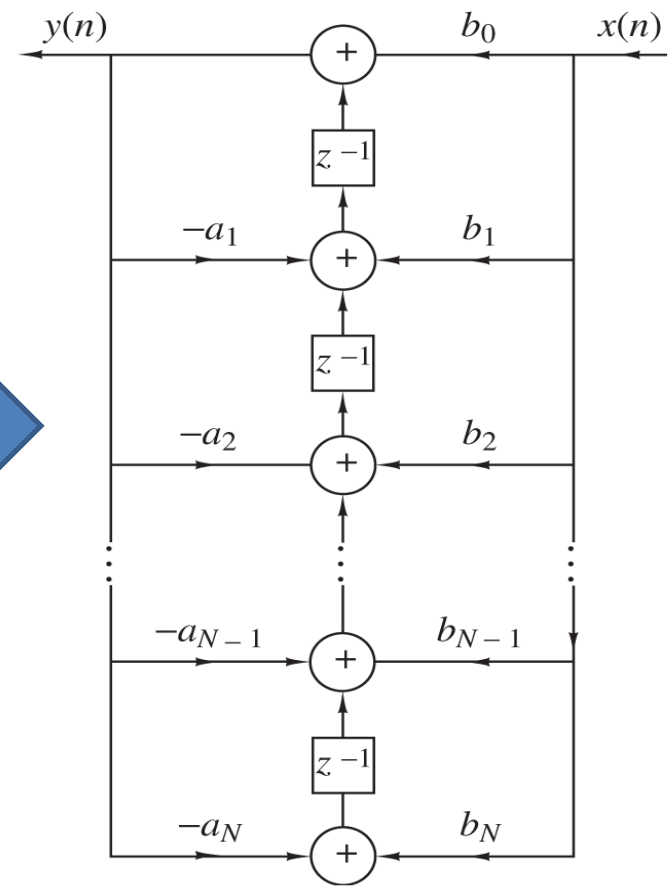
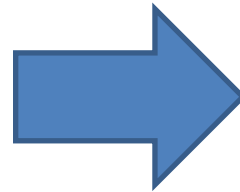
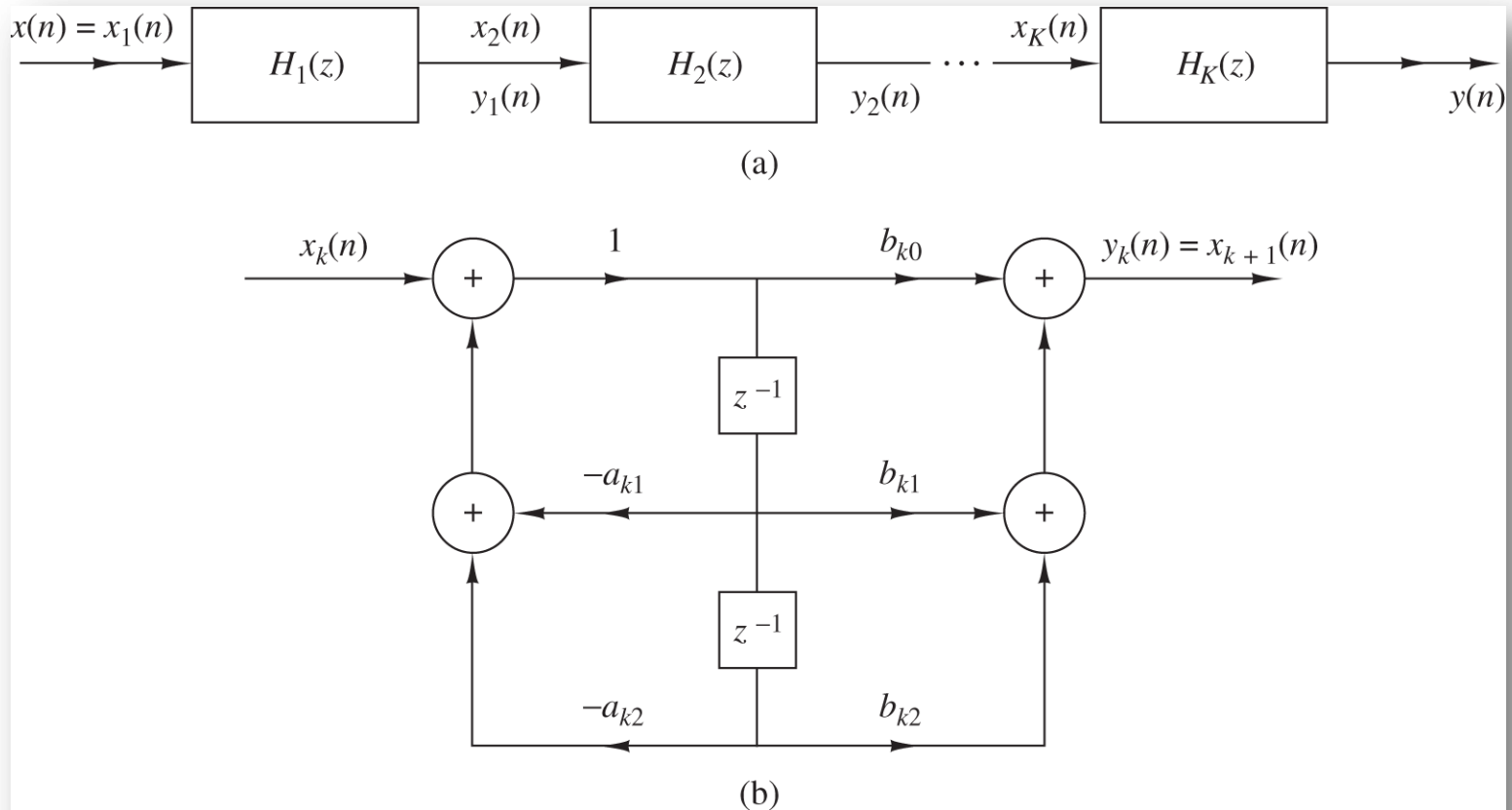


Figure 9.3.5 Transposed direct form II structure.

# Cascade-Form

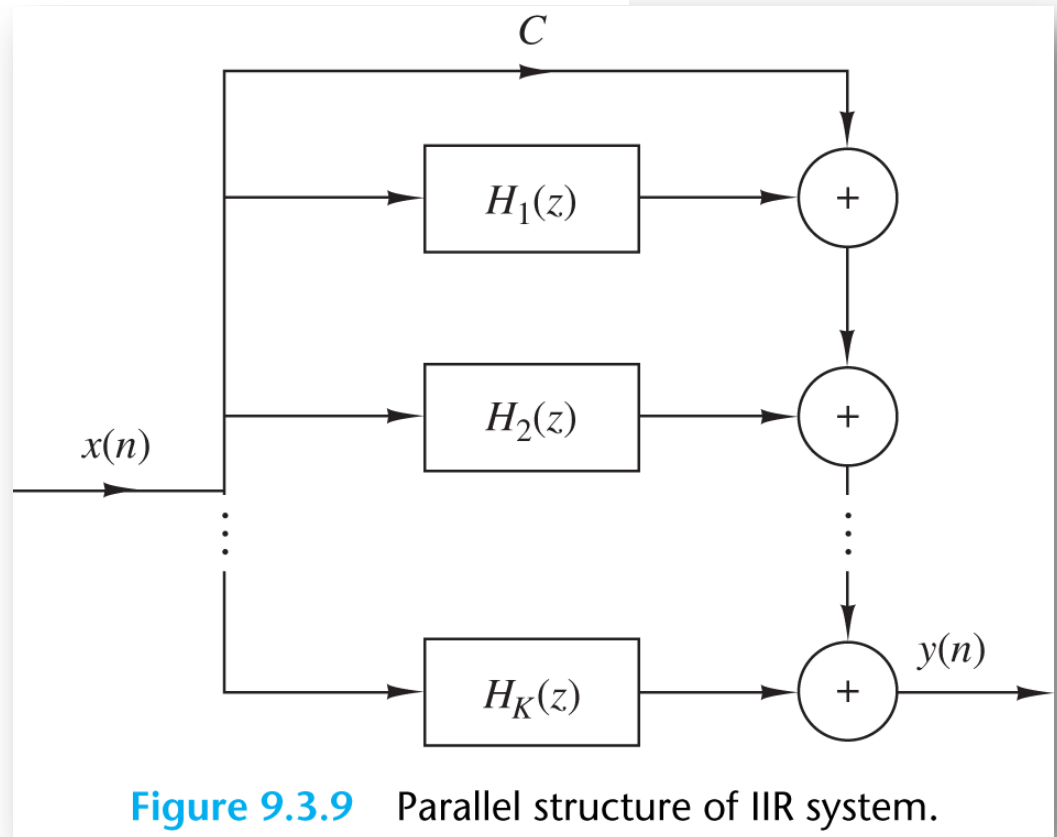


**Figure 9.3.8** Cascade structure of second-order systems and a realization of each second-order section.

# Parallel-Form

Performing partial fraction expansion of  $H(z)$ , we obtain

$$H(z) = C + \sum_{k=1}^N \frac{A_k}{a - p_k z^{-1}}$$



**Figure 9.3.9** Parallel structure of IIR system.